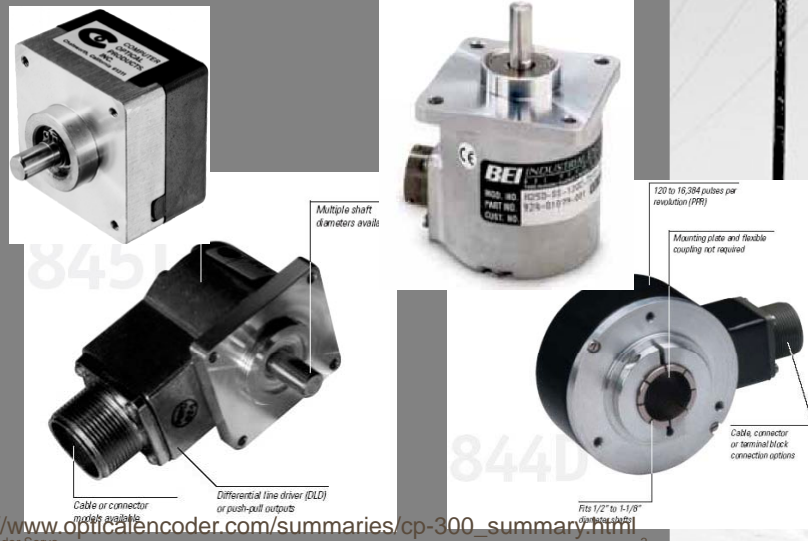


Various Forms of Encoders



http://www.opticalencoder.com/summaries/cp-300_summary.html

What is Coding

- (Not coding theorem)
- Digitally representing angular positions

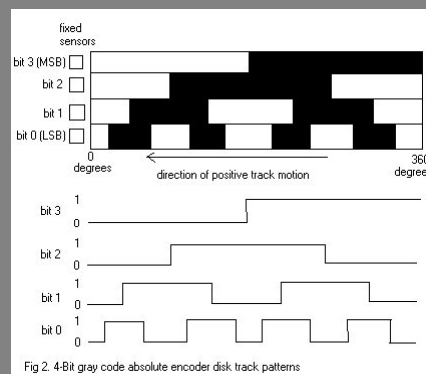


Fig 2. 4-Bit gray code absolute encoder disk track patterns

http://mechatronics.mech.northwestern.edu/design_ref/sensors/encoders.html

Gray vs. Binary Coding (4 bits)

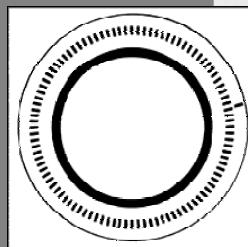
Decimal code	Rotation range (deg.)	Binary code	Gray code
0	0-22.5	0000	0000
1	22.5-45	0001	0001
2	45-67.5	0010	0011
3	67.5-90	0011	0010
4	90-112.5	0100	0110
5	112.5-135	0101	0111
6	135-157.5	0110	0101
7	157.5-180	0111	0100
8	180-202.5	1000	1100
9	202.5-225	1001	1101
10	225-247.5	1010	1111
11	247.5-270	1011	1110
12	270-292.5	1100	1010
13	292.5-315	1101	1011
14	315-337.5	1110	1001
15	337.5-360	1111	1000

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Absolute/Incremental Encoder

- **Absolute Encoder**
- **Incremental Encoder**



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SW Analysis

- **Strength**
 - Absolute position
 - Direct read out
- **Weakness**
 - Requires very high precision
 - Hard to manufacture
 - Easily corrupted data
 - Long parallel reading

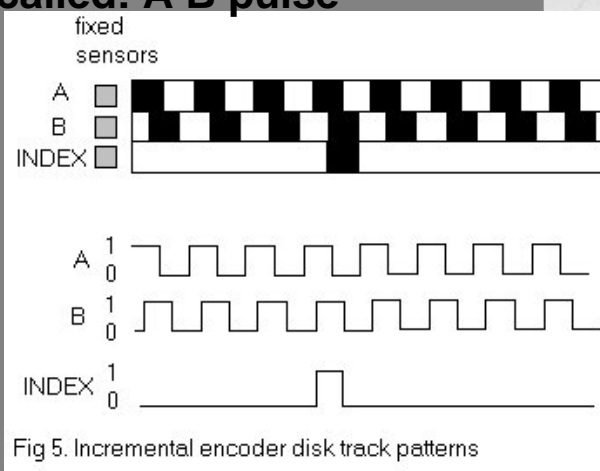
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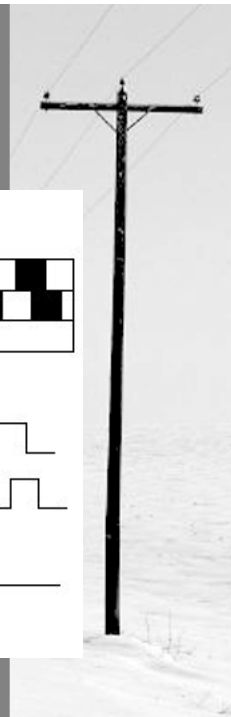


Incremental Encoder

- **Often called: A B pulse**

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To Detect Rotating Direction

- **90° offset for direction**
- **Forward direction**



- **When there is a reverse**



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SW Analysis

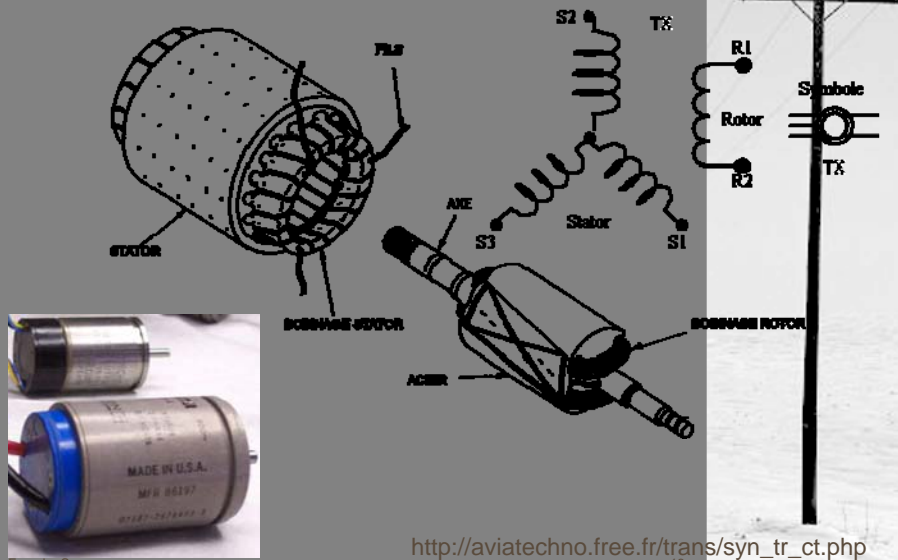
- **Strength**
 - Easy to manufacture
 - No limit on the rotation angle
 - High resolution
 - Sub-division for enhanced resolution
- **Weakness**
 - Require decoder
 - Serial (incremental) measurement

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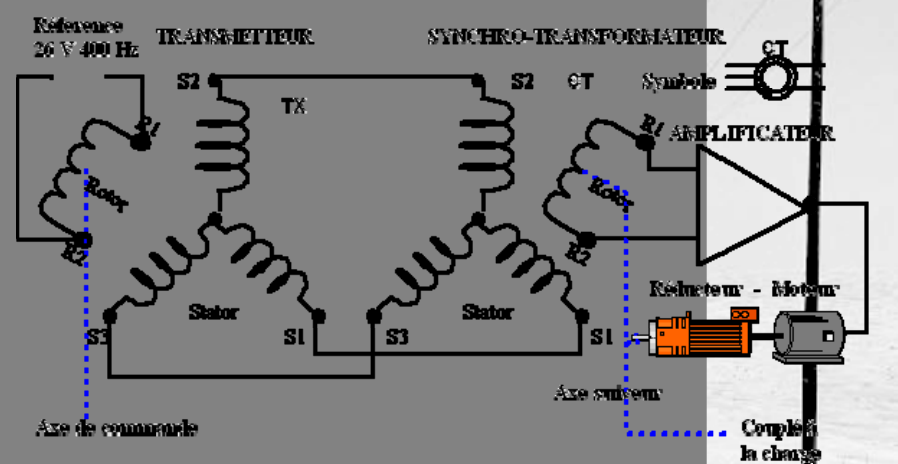
Synchro



http://aviatechno.free.fr/trans/syn_tr_ct.php

13

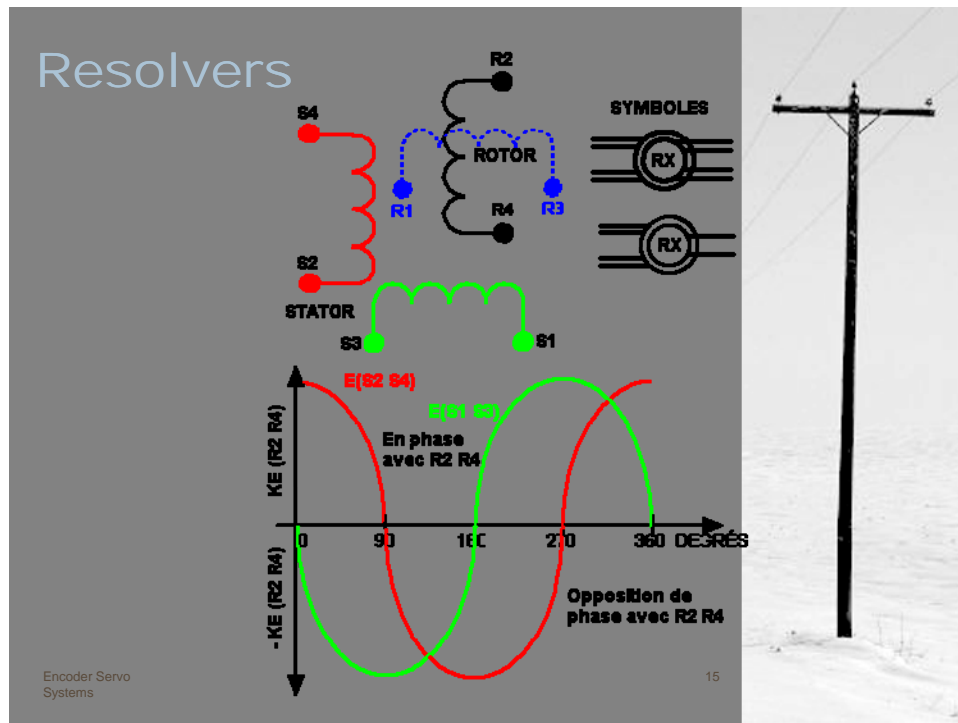
Transmission Synchro



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http://aviatechno.free.fr/trans/syn_tr_ct.php

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COMBINATIONAL LOGIC

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Basic Combinational Logic

- **Based on Boolean algebra and Karnaugh maps**
- **Use logic variables to represent signals**

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- **Example:**

- Take the air-conditioning control as an example:

A = temp too high,

B = humidity too high,

C = air-conditioning on,

- One can then say

$$C = A + B$$

- For air-conditioner control logic

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Boolean Logic

- There is a complete logic
- Upon the field of 0 and 1 (all the elements are 0 and 1)
- “+” stands for “or”
 - $A + 0 = A$
 - $A + A = A$
 - $A + 1 = 1$
 - $A + (-A) = 1$
- “.” stands for “and”
 - $A \bullet 1 = A$
 - $A \bullet A = A$
 - $A \bullet 0 = 0$
 - $A \bullet (-A) = 0$

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Algebraic Laws

- **Commutative**
 - $A + B = B + A$
 - $AB = BA$
- **Associative**
 - $A + (B + C) = (A + B) + C$
 - $A(BC) = (AB)C$
- **Distributive**
 - $A(B + C) = AB + AC$

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Basic Laws

- **Negative of negative becomes positive**
 - $A = \overline{\overline{A}}$
 - $\overline{AB} = \overline{A}\overline{B}$
 - $\overline{A+B} = \overline{A} + \overline{B}$
- **Conjugate**
 - For all Boolean equation
 - Change “or” to “and”
 - Change “and” to “or”
 - Switch “0” and “1”
 - We get the conjugate equation

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- **Example**

$$A + 0 = A$$

$$A \cdot 1 = A$$

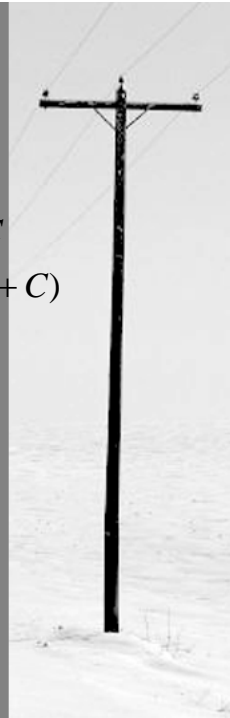
- **Example**

$$A(B + C) = AB + AC$$

$$A + BC = (A + B)(A + C)$$

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Controller Design

- “Sum of products” method
- Product of two variables: $Y=AB$
 - Possible combinations

A	B	Fundamental product
0	0	$\overline{A}\overline{B}$
0	1	$\overline{A}B$
1	0	$A\overline{B}$
1	1	AB

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- **Product of three variables: $Y = ABC$**
 - Possible combinations are

A	B	C	Fundamental product
0	0	0	$\overline{A}\overline{B}\overline{C}$
0	0	1	$\overline{A}\overline{B}C$
0	1	0	$\overline{A}B\overline{C}$
0	1	1	$\overline{A}BC$
1	0	0	$A\overline{B}\overline{C}$
1	0	1	$A\overline{B}C$
1	1	0	$AB\overline{C}$
1	1	1	ABC

- **And product of four variables, ...**
 - Possible combinations are

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To Establish The Required Logic

- **Example**
 - The required “truth table”

A	B	C	Y
0	0	0	0
0	0	1	0
0	1	0	$1 \rightarrow \overline{A}B\overline{C}$
0	1	1	0
1	0	0	0
1	0	1	$1 \rightarrow A\overline{B}C$
1	1	0	$1 \rightarrow AB\overline{C}$
1	1	1	$1 \rightarrow ABC$

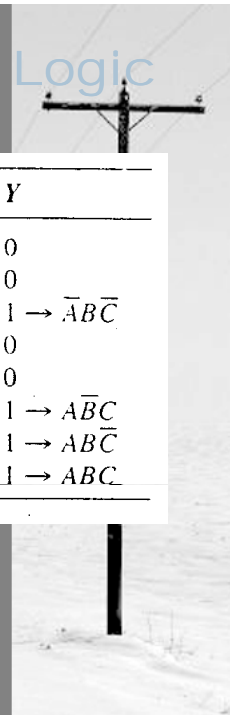
- The required actions becomes

$$Y = \overline{A}B\overline{C} + A\overline{B}C + AB\overline{C} + ABC$$

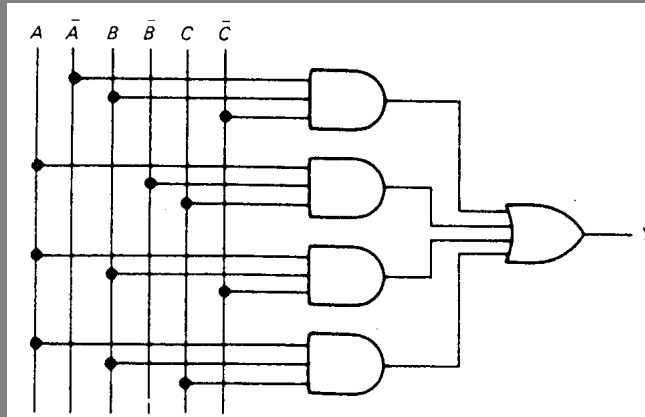
- Sum up all the actions!!

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- The circuit becomes



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Simplifying

- “Gate count” is a very important parameter defining logic circuit efficiency
- Simplifying reduces “Gate Count”

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Simplify Methods

- **Factoring**

- **Example**

$$\begin{aligned} Y &= A\bar{B} + AB \\ &= A(\bar{B} + B) \\ &= A \cdot 1 = A \end{aligned}$$

- **Example**

$$\begin{aligned} Y &= AB + AC + BD + CD \\ &= A(B + C) + D(B + C) \\ &= (A + D)(B + C) \end{aligned}$$

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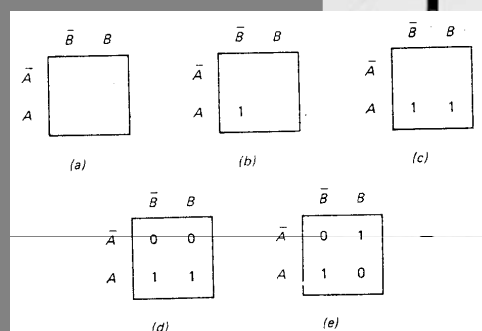
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Karnaugh Map

- **A effective tool to simplify a Boolean logic**
- **Construct a Karnaugh Map**
- **Two variables**
- **Mark all the “1’s”**

A	B	Y
0	0	0
0	1	0
1	0	1
1	1	1

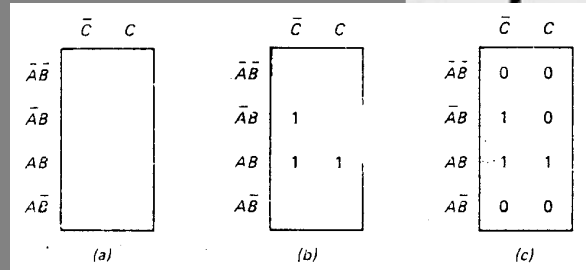
A	B	Y
0	0	0
0	1	1
1	0	1
1	1	0

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- Three variables
- Truth table

A	B	C	Y
0	0	0	0
0	0	1	0
0	1	0	1
0	1	1	0
1	0	0	0
1	0	1	0
1	1	0	1
1	1	1	1

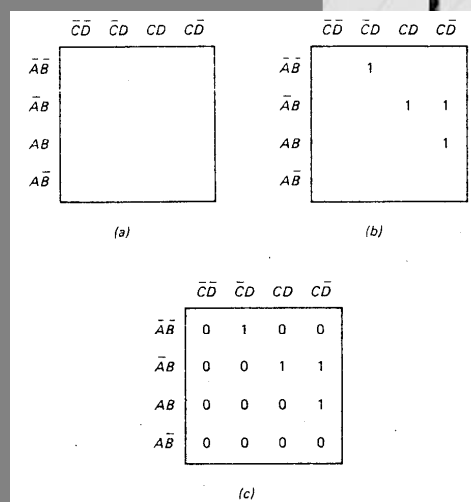


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- Four variables
- Truth table

A	B	C	D	Y
0	0	0	0	0
0	0	0	1	1
0	0	1	0	0
0	0	1	1	0
0	1	0	0	0
0	1	0	1	0
0	1	1	0	1
0	1	1	1	1
1	0	0	0	0
1	0	0	1	0
1	0	1	1	0
1	1	0	0	0
1	1	0	1	0
1	1	1	0	1
1	1	1	1	0



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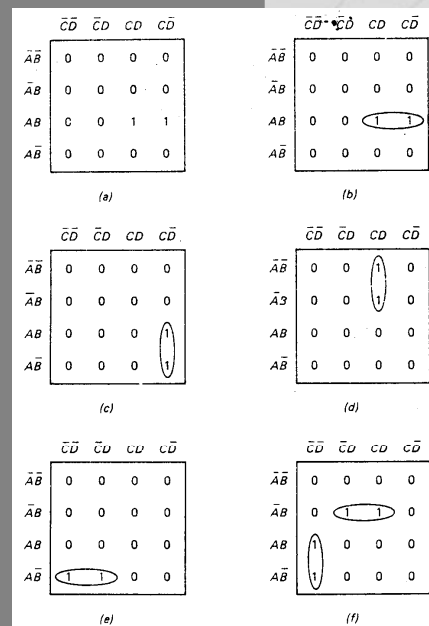
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- Find features

- Pair

$$\begin{aligned}
 Y &= ABCD + ABC\bar{D} \\
 &= ABC(D + \bar{D}) \\
 &= ABC
 \end{aligned}$$

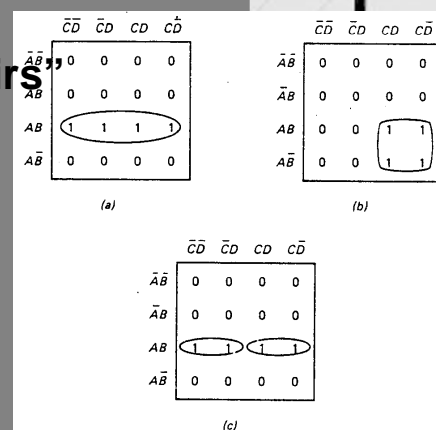
- Pairs are eliminated



- Quads can be eliminated

- “Quads” are the same as two “pairs”

$$\begin{aligned}
 Y &= AB\bar{C} + ABC \\
 &= AB(\bar{C} + C) \\
 &= AB
 \end{aligned}$$



- Octal:
- Like two “quads”

$$\begin{aligned}
 Y &= A\bar{C} + AC \\
 &= A(\bar{C} + C) \\
 &= A
 \end{aligned}$$

	$\bar{C}\bar{D}$	$\bar{C}D$	CD	$C\bar{D}$
$\bar{A}\bar{B}$	0	0	0	0
$\bar{A}B$	0	0	0	0
AB	1	1	1	1
$A\bar{B}$	1	1	1	1

(a)

	$\bar{C}\bar{D}$	$\bar{C}D$	CD	$C\bar{D}$
$\bar{A}\bar{B}$	0	0	0	0
$\bar{A}B$	0	0	0	0
AB	1	1	1	1
$A\bar{B}$	1	1	1	1

(b)

- Circle all the “1’s”

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- Note:
- Overlapping features
 - The equation in (b)

$$Y = A + \bar{A}B\bar{C}D$$

- But for (a)

$$Y = A + B\bar{C}D$$

	$\bar{C}\bar{D}$	$\bar{C}D$	CD	$C\bar{D}$
$\bar{A}\bar{B}$	0	0	0	0
$\bar{A}B$	0	1	0	0
AB	1	1	1	1
$A\bar{B}$	1	1	1	1

(a)

	$\bar{C}\bar{D}$	$\bar{C}D$	CD	$C\bar{D}$
$\bar{A}\bar{B}$	0	0	0	0
$\bar{A}B$	0	1	0	0
AB	1	1	1	1
$A\bar{B}$	1	1	1	1

(b)

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- **Roll over:**
 - The equation in (C)

$$Y = B\overline{C}\overline{D} + BCD$$

- But for (d)

$$Y = B\overline{D}$$

	$\overline{C}\overline{D}$	$\overline{C}D$	CD	$C\overline{D}$
$\overline{A}\overline{B}$	0	0	0	0
$\overline{A}B$	1	0	0	1
AB	1	0	0	1
$A\overline{B}$	0	0	0	0

	$\overline{C}\overline{D}$	$\overline{C}D$	CD	$C\overline{D}$
$\overline{A}\overline{B}$	0	0	0	0
$\overline{A}B$	1	0	0	1
AB	1	0	0	1
$A\overline{B}$	0	0	0	0

(c)
(d)

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- **Redundant features**
 - There is no need to repeat the covered features

	$\overline{C}\overline{D}$	$\overline{C}D$	CD	$C\overline{D}$
$\overline{A}\overline{B}$	0	0	0	0
$\overline{A}B$	0	1	0	0
AB	0	1	1	0
$A\overline{B}$	0	0	1	0

	$\overline{C}\overline{D}$	$\overline{C}D$	CD	$C\overline{D}$
$\overline{A}\overline{B}$	0	0	0	0
$\overline{A}B$	0	1	0	0
AB	0	1	1	0
$A\overline{B}$	0	0	1	0

(b)

- The Boolean equation

$$Y = B\overline{C}D + \overline{A}BD + ACD$$

$$= B\overline{C}D + ACD$$

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- Don't care states
 - Help further simplification

A	B	C	D	Y
0	0	0	0	1
0	0	0	1	0
0	0	1	0	0
0	0	1	1	1
0	1	0	0	1
0	1	0	1	1
0	1	1	0	0
0	1	1	1	1
1	0	0	0	X
1	0	0	1	X
1	0	1	0	X
1	0	1	1	X
1	1	0	0	X
1	1	0	1	X
1	1	1	0	X
1	1	1	1	X

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- Two ways to draw features

	CD	CD	CD	CD
$\bar{A}\bar{B}$	1	0	1	0
$\bar{A}B$	1	1	1	0
$A\bar{B}$	X	X	X	X
AB	X	X	X	X

	$\bar{C}\bar{D}$	$\bar{C}D$	CD	$C\bar{D}$
$\bar{A}\bar{B}$	1	0	1	0
$\bar{A}B$	1	1	1	0
$A\bar{B}$	X	X	X	X
AB	X	X	X	X

- The equation in (a)

$$Y = \bar{A}\bar{B}\bar{C}\bar{D} + \bar{A}\bar{B}\bar{C}D + \bar{A}\bar{B}C\bar{D} + \bar{A}\bar{B}CD + A\bar{B}\bar{C}\bar{D} + A\bar{B}\bar{C}D + A\bar{B}C\bar{D} + A\bar{B}CD$$

- The equation in (b)

$$Y = BD + \bar{C}\bar{D} + CD$$

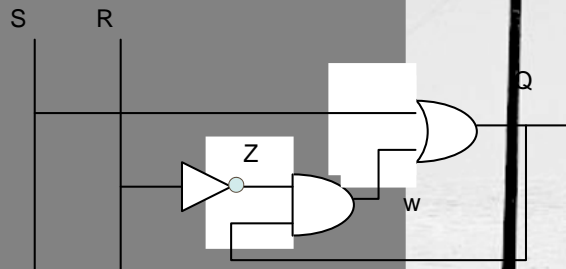
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Dynamic Sequential Logic Control

- Memory in Logic Circuits
- S-R Flip-flop:

- $Q^+ = S + \bar{R}Q$
 - 「or in "C", 」
 - $q = s \parallel (!r \ \$\$ q)$



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- The truth table

S	R	\bar{R}	Q	Q^+
0	0	1	0	0
1	0	1	0	1
0	0	1	1	1
0	1	0	1	0
0	0	1	0	0

- Q goes to 1 when S = 1, and remain 1 even if S goes to 0,
- Q goes to zero as R = 1, and remain 0 even if R goes to 0.

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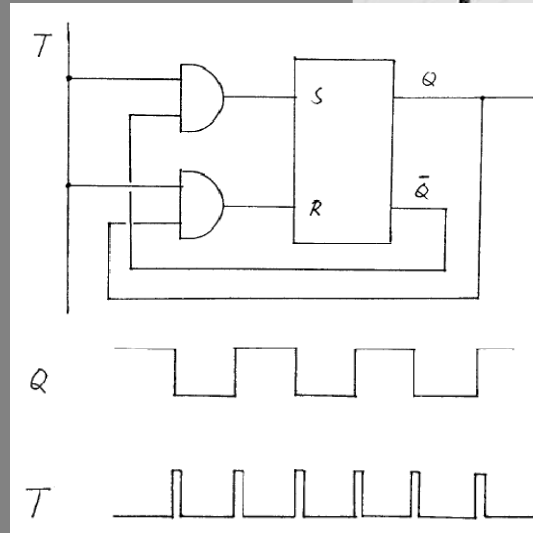
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- **T Flip-flop (Triggered flip-flop):**

- From S-R flip-flop and take

$$S = T\bar{Q} \quad R = TQ$$

- The output



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Design A Decoder

- **Two modes of operation**

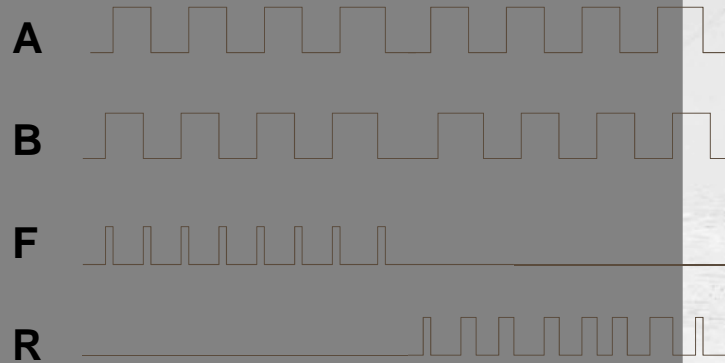
- Clocked circuit
- Asynchronous circuit
 - Pulsed mode
 - Systems in which the input is removed before any feedback signals have had a chance to change
 - Fundamental (Level) mode
 - Systems in which the input must remain without change until all of the changes in internal and feedback variables are complete.

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Quadrature Decoder

- **The desired output**



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Primitive Flow Table

- **A special form of next state table**

- Assume enough time must elapse after an input variable change for the system to reach a stable state
- Because the input remain applied to the system, the calculation of the next state must take both the current state and the input into account.
- The combination of the current and the input is called the “total state”
- A state is stable if the next state is equal to the current state

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- The primitive flow table is started by establishing a column of each possible input combination and a column for each of the outputs

The quadrant

B \ A	0	1
0	1	4
1	2	3

- Define states

- For forward motion, from S_1 proceed to the second quadrant will cause B: $0 \rightarrow 1$. The immediate change in input will cause a transition in column $1 \rightarrow 2$ and reach the unstable state S_2 .

- The outputs associated with the states are given in FR column

- For forward motion
 $S_1 \rightarrow S_2 \rightarrow S_3 \rightarrow S_4$,
 Reverse
 $S_5 \rightarrow S_6 \rightarrow S_7 \rightarrow S_8$

Quadrant	1	2	3	4	FR
A B	0 0	0 1	1 1	1 0	
S_1	S_2	S_3	S_4	0 0	
S_8	S_7	S_6	S_5	1 0	
S_1	S_2	S_3	S_4	0 0	
S_8	S_7	S_6	S_5	1 0	
S_1	S_2	S_3	S_4	0 1	
S_8	S_7	S_6	S_5	0 0	
S_1	S_2	S_3	S_4	0 1	
S_8	S_7	S_6	S_5	1 1	
S_1	S_2	S_3	S_4	1 1	
S_8	S_7	S_6	S_5		

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- Now go back and fill in the empty states.

- Moving in the reverse direction from S_2 requires a stable state with an output of 11, which is satisfied by state S_8 .
- Existing stable states can be used for these transitions until S_4 .
- Then there is no stable state already in the column for quadrant 3 that has the proper output, so the new state S_9 must be created.

- The primitive flow table is complete when all “legal” transitions have been filled in. The remaining transitions will become don't-cares.

Quadrant	1	2	3	4	FR
A B	0 0	0 1	1 1	1 0	
S_1	S_2		S_5	0 0	
S_8	S_7	S_3		1 0	
		S_2	S_4	0 0	
S_1		S_7	S_6	1 0	
S_8		S_6	S_5	0 1	
	S_7	S_4	S_4	0 0	
S_1	S_7	S_9		0 1	
S_8	S_2	S_9	S_4	1 1	
		S_5	S_5	1 1	

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Flow Table Reduction and Merger

- The primitive table needs reduction
- Stage 1:
 - Combine rows for which the stable states appear in the same column, and the outputs associated with each of the candidate rows must be the same and their next states are the same or equivalent.
 - Rows 3 and 6

Quadrant A B	1 0 0	2 0 1	3 1 1	4 1 0	FR
	s_1 s_8	s_2 s_2 s_7	s_3 s_3 s_7 s_9	s_5 s_4 s_4 s_4	0 0 1 0 0 0 1 0 0 1 0 0 0 1 1 1 1 1
	s_1 s_8	s_7 s_7 s_2	s_9 s_6 s_9	s_4 s_5 s_4 s_5	1 0 0 1 0 0 1 1 1 1

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- Stage 2:
 - Merge rows and allow several stable states to coexist on a single row.

	Next State				FR			
	AB	0 0	0 1	1 1	1 0	0 0	0 1	1 1
1	s_1	s_2	-	s_5	0 0	-	-	-
2	s_8	s_2	s_3	-	-	1 0	-	-
3	-	s_7	s_3	s_4	-	-	0 0	-
4	s_1	-	s_9	s_4	-	-	-	1 0
5	s_8	-	s_3	s_5	-	-	-	0 1
7	s_1	s_7	s_9	-	-	0 1	-	-
8	s_8	s_7	-	s_4	1 1	-	-	-
9	-	s_2	s_9	s_5	-	-	1 1	-

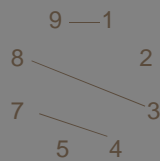
- The ones filled in are the ones associated with stable states; the rest have been left as don't cares for the moment.

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- **Merging rows:**

- Look for rows that do not have stable states in the same column, where the transitions in the candidate rows must either be to the same state, or one of the transitions must be a don't care



	Next State				FR				
	AB	00	01	11	10	00	01	11	10
1	s_1	s_2	-	s_6	00	-	-	-	-
2	s_8	s_7	s_3	-	-	10	-	-	-
3	-	s_7	s_3	s_4	-	-	00	-	-
4	s_1	-	s_9	s_2	-	-	-	10	-
5	s_8	-	s_3	s_6	-	-	-	-	01
7	s_1	s_7	s_9	-	-	01	-	-	-
8	s_8	s_7	-	s_4	11	-	-	-	-
9	-	s_2	s_9	s_5	-	-	11	-	-

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- **There is not unique choice of mergers**

	Next state				FR			
	A B				A B			
	00	01	11	10	00	01	11	10
a	1	2	9	5	00	--	11	--
b	8	2	3	-	--	10	--	--
c	8	2	3	-	--	10	--	--
d	1	7	9	4	11	--	00	--
e	8	-	3	5	--	--	--	01

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- **State assignment:**

- State assignment is not arbitrary:

- Example: 5 states $\Rightarrow [\log_2 5] = 3$
 - We need three bits to represent 5 states

		$Q_1+Q_2+Q_3$			
		A B			
$Q_1Q_2Q_3$		00	01	11	10
a 000	000	001	000	100	
b 001	010	001	010	-	
c 010	010	011	010	011	
d 011	000	011	000	011	
e 100	010	-	010	100	

- There will be trouble going from 011 to 000, might end up in 010 (stable).
 - This is called “race”.

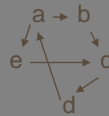
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- **To avoid “race”**

- Use all transitions are made with only one state change

- **Diagram of all possible transitions**



- **Use Karnaugh map**

- Make all transitions adjacent to each other. (Notice, augment states can be used to aid in the selection)

		Q_1Q_2			
		00	01	11	10
Q_3	0	a	b	(g)	
	1	d	c	e	(f)

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- New assignment of next state table

	$Q_1Q_2Q_3$	Next state				FR			
		A B				A B			
		00	01	11	10	00	01	11	10
a	000	000	010	000	100	00	10	11	01
b	010	011	010	011	-	1-	10	-0	--
c	011	011	001	011	001	11	01	00	10
d	001	000	001	000	001	00	01	11	10
e	111	011	-	011	111	-1	--	0-	01
(f)	101	-	-	-	111	--	--	--	01
(g)	100	-	-	-	101	--	--	--	01

- Output Assignment

- The output assignment are only associated with stable states.
 - Assign outputs in such a way that a transition through that unstable state will not cause a spurious pulse

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Exercise

- Find the register that holds the up/down count in the interface card in our system.
- Determine what kind of servo motor is used in our lab.
- Hook up the servo motor and try to read the up/down pulse count from the interface card.

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